

# Closed-Form Green's Functions for General Sources and Stratified Media

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**Abstract**—The closed-form Green's functions of the vector and scalar potentials in the spatial domain are presented for the sources of horizontal electric, magnetic, and vertical electric, magnetic dipoles embedded in general, multilayer, planar media. First, the spectral domain Green's functions in an arbitrary layer are derived analytically from the Green's functions in the source layer by using a recursive algorithm. Then, the spatial domain Green's functions are obtained by adding the contributions of the direct terms, surface waves, and complex images approximated by the Generalized Pencil of Functions Method (GPOF). In the derivations, the main emphasis is to put these closed-form representations in a suitable form for the solution of the mixed potential integral equation (MPIE) by the method of moments in a general three-dimensional geometry. The contributions of this paper are: 1) providing the complete set of closed-form Green's functions in spectral and spatial domains for general stratified media; 2) using the GPOF method, which is more robust and less noise sensitive, in the derivation of the closed-form spatial domain Green's functions; and 3) casting the closed-form Green's functions in a form to provide efficient applications of the method of moments.

## I. INTRODUCTION

**D**UE to the increased use of multilayer microstrip geometries in the application of microstrip antennas [1]–[6] and monolithic microwave integrated circuits [7]–[11], the layered geometries have recently attracted widespread attention. Therefore, a considerable amount of interest has been focused on the development of a rigorous and yet computationally efficient computer-aided design tools for microstrip geometries in a layered medium.

The rigorous analysis of layered microstrip structures requires the computation of the Green's functions for multilayer media, which are traditionally represented by the Sommerfeld integrals in the spatial domain, and by closed-form expressions in the spectral domain. When these traditional expressions for the Green's functions are employed in the method of moments (MoM), the numerical evaluation of the MoM matrix elements becomes very time consuming in either domain, because the integrals involved are oscillatory and slow decaying functions [12]. To alleviate this problem, the spatial domain Green's functions for the vector and scalar potentials, represented by the Sommerfeld integrals, are approximated by closed-form

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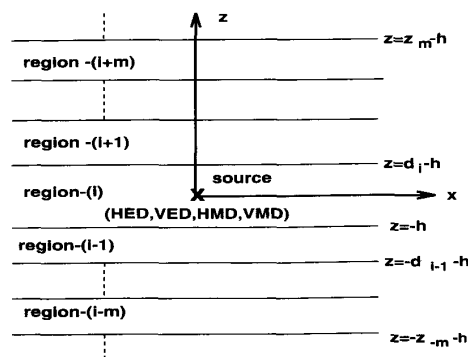


Fig. 1. Sources embedded in a multilayer medium.

expressions and used in the solution of the mixed potential integral equation (MPIE) by the MoM. This approach of approximating the spectral domain Green's functions was first proposed in [13] for a horizontal electric dipole (HED) over a thick substrate backed by a ground plane and extended to a geometry with a substrate and a superstrate with arbitrary thicknesses [14], using the original and least-square Prony's methods [16], respectively. It was demonstrated that the use of the closed-form Green's functions in the analysis of a microstrip geometry via the MoM improves the computational efficiency significantly [15].

In this paper, the closed-form Green's functions of the vector and scalar potentials of a Horizontal Electric Dipole (HED), Horizontal Magnetic Dipole (HMD), Vertical Electric Dipole (VED), and a Vertical Magnetic Dipole (VMD) located in an arbitrary layer of a planar-layered medium are presented. The layers are considered to have different dielectric and magnetic properties ( $\epsilon_r, \mu_r$ ) or are made of perfect electric or magnetic conductors (PEC, PMC), as shown in Fig. 1. The Green's functions are first obtained in the spectral domain, which can be represented in closed-form, in the source layer and these expressions are extended to an arbitrary layer using an iterative algorithm [8] for TE and TM components individually. Then, the spatial domain closed-form Green's functions are obtained by adding the contributions of the direct terms and surface waves to the complex images approximated by the Generalized Pencil of Functions (GPOF) method [17], which is less noise sensitive and more efficient than the Prony's methods. Since the spectral domain Green's functions, excluding the direct term and the surface wave components, are approximated by the GPOF method, their dependence on  $z$  must be kept in

explicit form to avoid the repeated use of GPOF approximation and to increase the computational efficiency in cases of vertical connections. Since the Green's functions for the vector and scalar potentials are not uniquely defined in stratified media [18], [19], the closed-form representations of an alternative formulation of the Green's functions are also provided, which might be used in cases where the vertical and horizontal sources are present at the same point [20].

Derivation of the closed-form Green's functions are given in Section II and the numerical considerations associated with their evaluation are discussed. In Section III, some numerical examples of the closed-form Green's functions are presented for two different multilayer geometries and the approximate Green's functions are compared with the exact Green's functions obtained by the numerical evaluation of the corresponding Sommerfeld integrals.

## II. FORMULATION

A general planar-layered medium is shown in Fig. 1. The source, (HED, HMD, VED or VMD) is embedded in region  $i$  and the observation point can be located in an arbitrary layer. Each layer can have different electric and magnetic properties ( $\epsilon_r, \mu_r$ ) and thickness ( $d_i$ ). The perfect electric or magnetic conducting planes and half space are also considered as layers for the formulation. The procedure for deriving the closed-form Green's functions can be summarized as the following steps:

- 1) Derivation of the Green's functions in the spectral domain.
  - a) Green's functions are derived in the source layer.
  - b) Green's functions in the observation layer are obtained using an iterative algorithm applied to each TE and TM component of the Green's functions in the source layer.
- 2) Derivation of the spatial domain, closed-form Green's functions.
  - a) Spectral Domain Green's functions, after having the surface wave poles and the direct terms extracted, are approximated in terms of complex exponentials obtained from the GPOF method.
  - b) Closed-form Green's functions are obtained analytically using the Sommerfeld identity for each complex exponential.

The derivation of the Green's functions for the vector and scalar potentials in the spectral domain follows the similar procedure given in [8], where the Green's functions for the electric and magnetic fields are obtained. Thus, the derived Green's functions, without giving the details of the derivation, are given in Section II-A with all the necessary definitions of the reflection coefficients and amplitudes for the sake of completeness and for later references.

All of the Green's functions presented here are for the vector and scalar potentials that are indeed not uniquely defined in stratified media [18], [19]. Therefore, different sets of Green's functions for the vector and scalar potentials can be chosen to satisfy the same boundary conditions. The following form

of the Green's function is commonly used and referred as the traditional form for the vector potentials

$$\bar{G}_{A,F} = (\hat{x}\hat{x} + \hat{y}\hat{y})G_{xx} + \hat{z}\hat{x}G_{zx} + \hat{z}\hat{y}G_{zy} + \hat{z}\hat{z}G_{zz} \quad (1)$$

and for the scalar potentials,  $G_{x,y}^{q_e,m}$  and  $G_z^{q_e,m}$  [20]. Note that in this representation, the scalar potentials of the point charges associated with the horizontal and vertical dipoles are not identical. This results in some difficulties in the solution of the mixed potential integral equation for a geometry where both the horizontal and vertical sources (HED and VED or HMD and VMD) are present at the same point, as in the case of a microstrip etch fed by a vertical probe. To overcome this difficulty, an alternative formulation is proposed in [19] and adopted in this paper to the procedure described above. The alternative representations of the Green's functions are given in the Appendix.

### A. Green's Functions in the Spectral Domain

To derive the spectral domain Green's function for the source layer (layer  $i$ ), the  $z$  dependence of the fields in the source region is written as the sum of the direct term and up- and down-going waves due to the reflections from the boundaries at  $z = -h$  and  $z = d_i - h$ , respectively. The coefficients of the up- and down-going waves can be obtained in terms of the generalized reflection coefficients by applying the appropriate boundary conditions. The spectral domain Green's functions (traditional form) in the source layer are obtained for the sources of HED, HMD, VED, and VMD as

HED

$$\tilde{G}_{xx}^A = \frac{\mu_i}{2jk_{z_i}} [e^{-jk_{z_i}|z|} + A_h^e e^{jk_{z_i}z} + C_h^e e^{-jk_{z_i}z}] \quad (2)$$

$$\tilde{G}_{zx}^A = \frac{-\mu_i}{2jk_{z_i}} \left[ \frac{k_x k_{z_i}}{k_\rho^2} (A_h^e + B_h^e) e^{jk_{z_i}z} + \frac{k_x k_{z_i}}{k_\rho^2} (D_h^e - C_h^e) e^{-jk_{z_i}z} \right] \quad (3)$$

$$\tilde{G}_x^{q_e} = \frac{1}{j2\epsilon_i k_{z_i}} \left[ e^{-jk_{z_i}|z|} + \frac{k_{z_i}^2 B_h^e + k_i^2 A_h^e}{k_\rho^2} e^{jk_{z_i}z} + \frac{k_i^2 C_h^e - k_{z_i}^2 D_h^e}{k_\rho^2} e^{-jk_{z_i}z} \right] \quad (4)$$

HMD

$$\tilde{G}_{xx}^F = \frac{\epsilon_i}{2jk_{z_i}} [e^{-jk_{z_i}|z|} + A_h^m e^{jk_{z_i}z} + C_h^m e^{-jk_{z_i}z}] \quad (5)$$

$$\tilde{G}_{zx}^F = \frac{-\epsilon_i}{2jk_{z_i}} \left[ \frac{k_x k_{z_i}}{k_\rho^2} (A_h^m + B_h^m) e^{jk_{z_i}z} + \frac{k_x k_{z_i}}{k_\rho^2} (D_h^m - C_h^m) e^{-jk_{z_i}z} \right] \quad (6)$$

$$\tilde{G}_x^{q_m} = \frac{1}{j2\mu_i k_{z_i}} \left[ e^{-jk_{z_i}|z|} + \frac{k_{z_i}^2 B_h^m + k_i^2 A_h^m}{k_\rho^2} e^{jk_{z_i}z} + \frac{k_i^2 C_h^m - k_{z_i}^2 D_h^m}{k_\rho^2} e^{-jk_{z_i}z} \right] \quad (7)$$

VED

$$\tilde{G}_{zz}^A = \frac{\mu_i}{j2k_{z_i}} [e^{-jk_{z_i}|z|} + A_v^e e^{-jk_{z_i}z} + B_v^e e^{jk_{z_i}z}] \quad (8)$$

$$\tilde{G}_z^{q_e} = \frac{1}{j2k_{z_i}\epsilon_i} [e^{-jk_{z_i}|z|} + C_v^e e^{-jk_{z_i}z} + D_v^e e^{jk_{z_i}z}] \quad (9)$$

VMD:

$$\tilde{G}_{zz}^F = \frac{\epsilon_i}{j2k_{z_i}} [e^{-jk_{z_i}|z|} + A_v^m e^{-jk_{z_i}z} + B_v^m e^{jk_{z_i}z}] \quad (10)$$

$$\tilde{G}_z^{q_m} = \frac{1}{j2k_{z_i}\mu_i} [e^{-jk_{z_i}|z|} + C_v^m e^{-jk_{z_i}z} + D_v^m e^{jk_{z_i}z}] \quad (11)$$

where  $\tilde{G}_{ij}^{A,F}$  denotes the spectral domain Green's functions for the vector potentials in direction- $i$  due to a unit  $j$ -directed current element,  $\tilde{G}_i^{q_e,m}$  represents the Green's function of the scalar potential in the spectral domain due to a unit  $i$ -directed electric or magnetic current element,  $k_i^2 = k_\rho^2 + k_{z_i}^2$ , the superscripts  $A$  and  $F$  represent the magnetic and the electric vector potentials, respectively,  $q_e$  and  $q_m$  represent electric and magnetic scalar potentials, respectively. The coefficients,  $A_{h,v}^{e,m}, B_{h,v}^{e,m}, C_{h,v}^{e,m}, D_{h,v}^{e,m}$  are functions of the generalized reflection coefficients  $\tilde{R}_{TE,TM}$  [8], and are given by

$$A_h^{e,m} = e^{-jk_{z_i}(d_i-h)} \tilde{R}_{TE,TM}^{i,i+1} [e^{-jk_{z_i}(d_i-h)} + \tilde{R}_{TE,TM}^{i,i-1} e^{-jk_{z_i}(d_i+h)}] M_i^{TE,TM} \quad (12)$$

$$B_h^{e,m} = e^{-jk_{z_i}(d_i-h)} \tilde{R}_{TM,TE}^{i,i+1} [e^{-jk_{z_i}(d_i-h)} - \tilde{R}_{TM,TE}^{i,i-1} e^{-jk_{z_i}(d_i+h)}] M_i^{TM,TE} \quad (13)$$

$$C_h^{e,m} = e^{-jk_{z_i}h} \tilde{R}_{TE,TM}^{i,i-1} [e^{-jk_{z_i}h} + \tilde{R}_{TE,TM}^{i,i+1} e^{-jk_{z_i}(2d_i-h)}] M_i^{TE,TM} \quad (14)$$

$$D_h^{e,m} = e^{-jk_{z_i}h} \tilde{R}_{TM,TE}^{i,i-1} [-e^{-jk_{z_i}h} + \tilde{R}_{TM,TE}^{i,i+1} e^{-jk_{z_i}(2d_i-h)}] M_i^{TM,TE} \quad (15)$$

$$A_v^{e,m} = e^{-jk_{z_i}h} \tilde{R}_{TM,TE}^{i,i-1} [e^{-jk_{z_i}h} + \tilde{R}_{TM,TE}^{i,i+1} e^{-jk_{z_i}(2d_i-h)}] M_i^{TM,TE} \quad (16)$$

$$B_v^{e,m} = e^{-jk_{z_i}(d_i-h)} \tilde{R}_{TM,TE}^{i,i+1} [e^{-jk_{z_i}(d_i-h)} + \tilde{R}_{TM,TE}^{i,i-1} e^{-jk_{z_i}(d_i+h)}] M_i^{TM,TE} \quad (17)$$

$$C_v^{e,m} = e^{-jk_{z_i}h} \tilde{R}_{TM,TE}^{i,i-1} [-e^{-jk_{z_i}h} + \tilde{R}_{TM,TE}^{i,i+1} e^{-jk_{z_i}(2d_i-h)}] M_i^{TM,TE} \quad (18)$$

$$D_v^{e,m} = e^{-jk_{z_i}(d_i-h)} \tilde{R}_{TM,TE}^{i,i+1} [-e^{-jk_{z_i}(d_i-h)} + \tilde{R}_{TM,TE}^{i,i-1} e^{-jk_{z_i}(d_i+h)}] M_i^{TM,TE} \quad (19)$$

where

$$M_i^{TE,TM} = [1 - \tilde{R}_{TE,TM}^{i,i+1} \tilde{R}_{TE,TM}^{i,i-1} e^{-jk_{z_i}2d_i}]^{-1} \quad (20)$$

$$\tilde{R}_{TE,TM}^{j+1,j} = \frac{R_{TE,TM}^{j+1,j} + \tilde{R}_{TE,TM}^{j,j-1} e^{-jk_{z_j}2d_j}}{1 - R_{j,j+1} \tilde{R}_{TE,TM}^{j,j-1} e^{-jk_{z_j}2d_j}} \quad (21)$$

and  $R$  and  $\tilde{R}$  are the Fresnel and generalized reflection coefficients [8] for which the subscripts TE and TM represent the polarization of the wave, and the superscripts  $(i, i-1)$  or  $(i, i+1)$  show the layer numbers. The subscripts  $h$  and  $v$  used in the coefficients (12)–(17) represent the orientation of the source, horizontal and vertical, respectively, while the superscripts  $e$  and  $m$  denote the type of the source, electric and magnetic, respectively. It should be noted that the horizontal Green's functions for the  $y$ -oriented dipoles can be obtained

simply by setting  $\tilde{G}_{yy}^{A,F} = \tilde{G}_{xx}^{A,F}$ ,  $\tilde{G}_{zy}^{A,F}/k_y = \tilde{G}_{zx}^{A,F}/k_x$ , and  $\tilde{G}_y^{q_e,m} = \tilde{G}_x^{q_e,m}$ .

The amplitudes of the up- and down-going waves in a layer different from the source layer are related to those in the adjacent layers by

$$A_j^- = A_{j+1}^- \frac{T_{j+1,j} e^{-j(k_{z_{j+1}} - k_{z_j})(h+z_{-m+1})}}{1 - R_{j,j+1} \tilde{R}_{j,j-1} e^{-jk_{z_j}2d_j}} \quad (22)$$

where  $A_j^-$  and  $A_{j+1}^-$  are the amplitudes of the down-going waves in layers  $j$  and  $j+1$ , respectively,  $(j = i-m)$ ,  $T$  is the transmission coefficient, and  $z_m$  is the distance between the lower boundary of the source layer  $i$  and the lower boundary of layer  $j$  (see Fig. 1). Similarly, the amplitudes of the up-going waves in layer  $j = i+m$  can be written as

$$A_j^+ = A_{j-1}^+ \frac{T_{j-1,j} e^{-j(k_{z_{j-1}} - k_{z_j})(z_{m-1} + d_i - h)}}{1 - R_{j,j-1} \tilde{R}_{j,j+1} e^{-jk_{z_j}2d_j}} \quad (23)$$

Therefore, starting from the source layer, the field expressions for any layer can be obtained iteratively [8].

### B. Closed-form Green's Functions in the Spatial Domain

The spatial domain Green's functions are represented by the Sommerfeld integral [21] as

$$G^{A,F,q_e,q_m} = \frac{1}{4\pi} \int_{SIP} dk_\rho k_\rho H_o^{(2)}(k_\rho \rho) \tilde{G}^{A,F,q_e,q_m}(k_\rho) \quad (24)$$

where  $G$  and  $\tilde{G}$  are the Green's functions in the spatial and spectral domains, respectively,  $H_o^{(2)}$  is the Hankel function of the second kind, and SIP is the Sommerfeld integration path. The Sommerfeld integral given in (24) cannot be integrated analytically, except for a few special cases. On the other hand, if the spectral domain representation of the Green's function,  $\tilde{G}$ , in the integrand can be approximated in terms of complex exponentials, then the analytical evaluation of the integral (24) becomes possible via the Sommerfeld identity. Since the contributions of the direct terms,  $e^{-jk_{z_i}|z|}/k_{z_i}$ , and the surface waves can be calculated analytically, they are excluded from the expressions to be approximated, [13], [14].

In approximating the spectral domain Green's functions, the GPOF method, which is based on solving a generalized eigenvalue problem, is used [17]. The spectral domain Green's functions,  $\tilde{G}$ , are uniformly sampled along an integration path,  $k_z = k[-jt + (1-t/T_o)]$  deformed from SIP [13] and approximated by complex exponentials as

$$\tilde{G} \cong \sum_{m=1}^N a_m e^{-b_m k_{z_i}} \quad (25)$$

where  $N$  is the number of exponentials used in the approximation. Then, the Sommerfeld identity

$$\frac{e^{-jkr}}{r} = -\frac{j}{2} \int_{SIP} dk_\rho k_\rho H_o^{(2)}(k_\rho \rho) \frac{e^{-jk_z|z|}}{k_z} \quad (26)$$

is employed to obtain the Green's functions in the following form, referred as the closed-form

$$G \cong \sum_{m=1}^N a_m \frac{e^{-jk_i r_m}}{r_m} + \text{direct term} + \text{surface waves} \quad (27)$$

where  $r_m = \sqrt{\rho^2 - b_m^2}$  represents a complex distance. Since the surface wave contributions for thin-layered structures are small, the exclusion of the surface wave contributions is not critical for such geometries; meanwhile, the extraction of the surface wave poles could improve the approximation for geometries with thick layers. Note that the approximation of the spectral domain Green's functions (2)–(11) must be performed for the terms in the square brackets, i.e., the terms are apart from  $1/k_{z_i}$  after extracting the direct term and surface wave poles, to be able to use the Sommerfeld identity. In addition,  $k_x$  and  $k_y$  parameters in  $\tilde{G}_{zx}^{A,F}$  and  $\tilde{G}_{zy}^{A,F}$ , respectively, are excluded in the approximation and their contributions are added in the spatial domain (after having obtained the spatial domain representations of  $\tilde{G}_{zx}^{A,F}/k_x$  and  $\tilde{G}_{zy}^{A,F}/k_y$ ) by differentiating analytically with respect to  $x$  and  $y$ , respectively.

### C. Closed-Form Green's Functions for MoM Applications

So far, a general procedure to obtain the closed-form Green's functions for the vector and scalar potentials has been explained, but nothing has been done yet to make these closed-form expressions numerically efficient when they are used in conjunction with the method of moments. As mentioned above, the terms in the square brackets in (2)–(11) are to be sampled uniformly along  $k_{z_i}$  and approximated by complex exponentials. To do so, one needs to fix the vertical coordinate variable  $z$ ; that is, the approximation technique GPOF has to be applied for each  $z$  value involved in the analysis. For cases of horizontal conductors only, this wouldn't cause computational inefficiency because the conductors must be placed at constant  $z$ -planes, resulting in the following MoM matrix element:

$$\langle T_{xm}, G_{xx}^A * B_{xn} \rangle + \frac{1}{\omega^2} \left\langle T_{xm}, \frac{\partial}{\partial x} \left[ G_x^{qe} * \frac{\partial B_{xn}}{\partial x} \right] \right\rangle \quad (28)$$

where  $T_{xm}$  and  $B_{xn}$  are testing and basis functions, respectively.  $G_{xx}^A$  and  $G_x^{qe}$  used in this formulation are approximated by GPOF for constant  $z$ 's corresponding to the planes of the conductors in the geometry. However, in geometries with both vertical and horizontal conductors with  $z$  and  $x$ -directed current components, respectively, typical MoM matrix elements can have the terms of

$$\begin{aligned} \langle T_{zm}, G_{zx}^A * B_{zn} \rangle, & \quad \left\langle T_{zm}, \frac{\partial}{\partial z} \left[ G_x^{qe} * \frac{\partial B_{zn}}{\partial x} \right] \right\rangle \\ \langle T_{zm}, G_{zz}^A * B_{zn} \rangle, & \quad \left\langle T_{zm}, \frac{\partial}{\partial z} \left[ G_z^{qe} * \frac{\partial B_{zn}}{\partial z} \right] \right\rangle \end{aligned} \quad (29)$$

where  $G_{zx}^A$ ,  $G_{zz}^A$ ,  $G_x^{qe}$  and  $G_z^{qe}$  need to be approximated at every observation point  $z$  and/or source point  $z'$  values (assuming the origin is at the bottom of the source layer for the application of the MoM, the coordinates used in the derivation of the Green's functions here can be transformed by  $z \leftarrow z - z'$ ,  $h \leftarrow z'$ ) in the integration due to the testing and expansion processes along a vertical conductor. This would defeat the purpose of using the closed-form Green's functions in a MoM application. To circumvent the problem associated with the testing process, which corresponds to integration along  $z$ , the

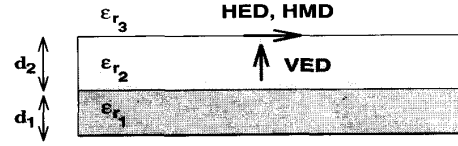


Fig. 2. Geometry of a 4-layer structure. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10$ ,  $d_1 = 0.075$  cm,  $\epsilon_{r2} = 2$ ,  $d_2 = 0.15$  cm,  $\epsilon_{r3} = 1$ ,  $h = d_2$  for HED and HMD,  $h = d_2/2$  for VED,  $z = 0.0$  cm.

GPOF method can be applied to the complex coefficients of  $e^{\pm jk_{z_i}z}$  in all the Green's functions except  $\tilde{G}_{xx}^{A,F}$ . Hence, the  $z$ -dependence in the closed-form Green's functions becomes explicit and the testing procedure along the  $z$ -direction can be performed analytically for some testing functions like uniform and roof-top functions, which further improves the computational efficiency of using the closed-form Green's functions in conjunction with the method of moments. Another technique, proposed here, to overcome the above mentioned difficulties is to interchange the order of integration (29), provided that the basis and testing functions are so chosen that the involved integrals are uniformly convergent [12], and carrying out integration over  $z$  analytically for spectral domain representation of the Green's function multiplied with the testing function. Next, the approximation method, GPOF, is applied to the resulting spectral domain function. For the inner product terms involving both the  $z$  and  $z'$  integrations, these integrals can be performed analytically by using the procedure described above and subsequently applying the GPOF algorithm. Note that the spectral representations of the Green's functions are exponential functions of  $z$  and  $z'$ , and this permits us to carry out  $z$  and  $z'$  integrations analytically.

### III. NUMERICAL RESULTS AND DISCUSSIONS

The closed-form Green's functions presented in Section II can be used for planar-layered geometries having an arbitrary number of layers with arbitrary layer parameters and general sources. In this section, a multilayer geometry is investigated for different types of sources and the Green's functions obtained using the closed-forms (approximate) are compared with the exact Green's functions, calculated by evaluating the corresponding Sommerfeld integrals numerically.

The geometry investigated in this paper consists of a substrate and a superstrate with three different dipoles (HED, HMD, and VED) and modeled as a 4-layer structure with the following parameters—layer-0: PEC, layer-3: half-space,  $\epsilon_{r1} = 10$ ,  $\epsilon_{r2} = 2$ ,  $\epsilon_{r3} = 1$ ,  $d_1 = 0.075$  cm,  $d_2 = 0.15$  cm, as shown in Fig. 2. The horizontal dipoles (HED and HMD) are located at the air-dielectric interface ( $h = d_2$ ) and the vertical dipole is located in the middle of the top layer ( $h = d_2/2$ ). In all three cases, the observation points are chosen at the source plane ( $z = z' = 0.0$  cm), which is the worst case as far as the convergence of the Green's functions are concerned. Figs. 3–9 show the magnitude of  $G_{xx}^A$ ,  $\int G_{zx}^A dx$ ,  $G_x^{qe}$ ,  $G_{xx}^F$ ,  $G_x^{qe}$ ,  $G_{zz}^A$  and  $G_z^{qm}$  with respect to the distance  $k_o\rho$ , which are obtained using both the closed-form representations and numerically evaluating the

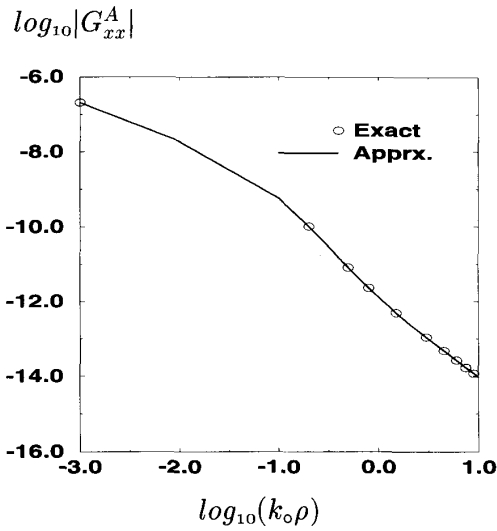


Fig. 3. Magnitude of the Green's function for the vector potential,  $G_{xx}^A$ , for an HED. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10, d_1 = 0.075$  cm,  $\epsilon_{r2} = 2, d_2 = 0.15$  cm,  $\epsilon_{r3} = 1, h = d_2, z = 0.0$  cm,  $f = 1$  GHz.

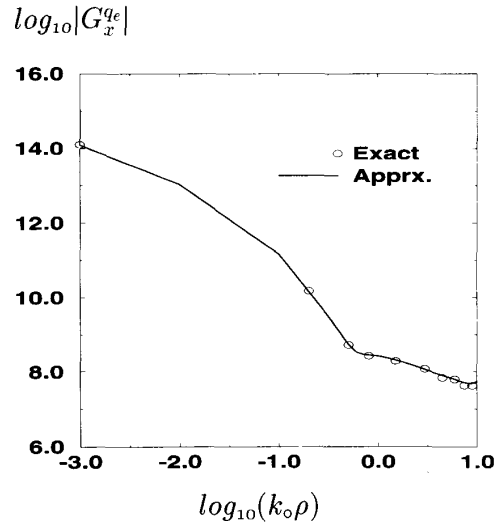


Fig. 5. Magnitude of the Green's function for the scalar potential,  $G_x^{qe}$ , for an HED. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10, d_1 = 0.075$  cm,  $\epsilon_{r2} = 2, d_2 = 0.15$  cm,  $\epsilon_{r3} = 1, h = d_2, z = 0.0$  cm,  $f = 1$  GHz.

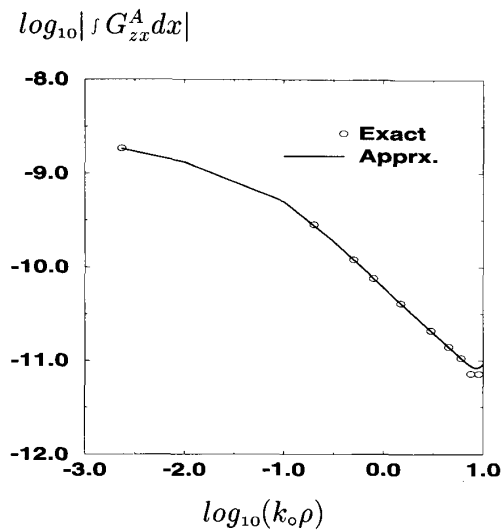


Fig. 4. Magnitude of the Green's function for the vector potential,  $\int G_{zx}^A dx$ , for an HED. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10, d_1 = 0.075$  cm,  $\epsilon_{r2} = 2, d_2 = 0.15$  cm,  $\epsilon_{r3} = 1, h = d_2, z = 0.0$  cm,  $f = 1$  GHz.

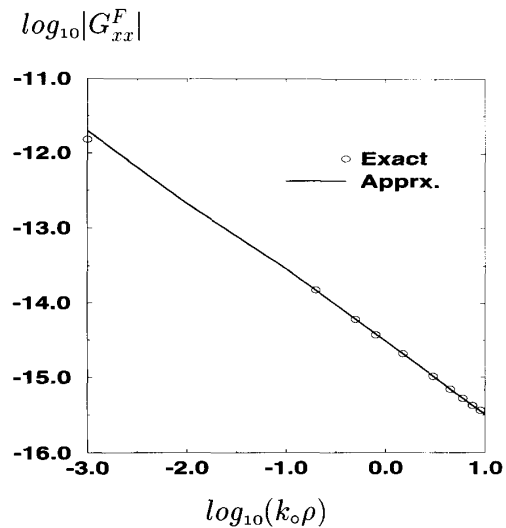


Fig. 6. Magnitude of the Green's function for the vector potential,  $G_{xx}^F$ , for an HMD. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10, d_1 = 0.075$  cm,  $\epsilon_{r2} = 2, d_2 = 0.15$  cm,  $\epsilon_{r3} = 1, h = d_2, z = 0.0$  cm,  $f = 1$  GHz.

corresponding Sommerfeld integrals at a frequency of 1 GHz. In Fig. 4  $|\int G_{zx}^A dx|$  is given, instead of  $|G_{zx}^A|$ , because the approximation is performed on  $\tilde{G}_{zx}^A/k_x$ . Fig. 10 shows the approximate and the exact Green's functions,  $G_{zz}^A$ , calculated using the alternative form.

In approximating the spectral domain Green's functions using the GPOF, the choice of the number of samples used to represent a Green's function in the spectral domain, the maximum sampled value of  $k_z$ , and the number of exponentials used to approximate the spectral domain Green's functions

depend on the parameters of the multilayer geometry, the type and the orientation of the dipole, and the Green's function to be approximated. For example, it is observed that fewer sampling points, a fewer number of exponentials, and a smaller value of  $T_o$  are required for  $G_{xx}^A$  than those required for  $G_x^{qe}$ . This is because the vector potential contributes to the far field in the spatial domain; the major contributions in the spectral domain come from the region close to the origin. On the other hand, the scalar potential, which contributes to the near field dominantly, extends to larger values of  $k_\rho$  in the spectral domain. As one of the contributions of this paper, the

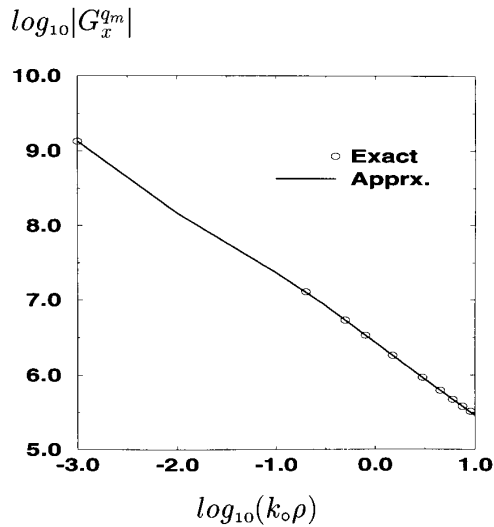


Fig. 7. Magnitude of the Green's function for the scalar potential,  $G_x^{qm}$ , for an HMD. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10$ ,  $d_1 = 0.075$  cm,  $\epsilon_{r2} = 2$ ,  $d_2 = 0.15$  cm,  $\epsilon_{r3} = 1$ ,  $h = d_2$ ,  $z = 0.0$  cm,  $f = 1$  GHz.

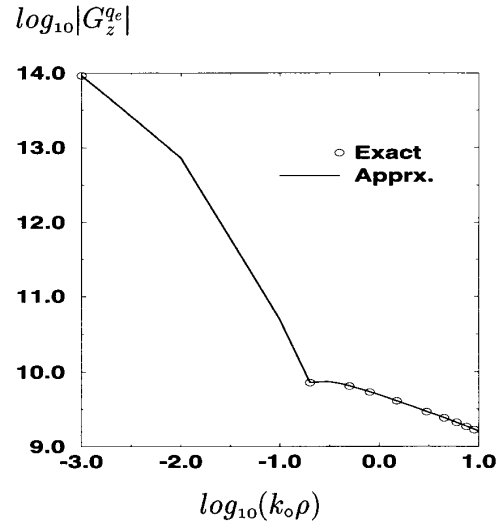


Fig. 9. Magnitude of the Green's function for the scalar potential,  $G_z^{qe}$ , for a VED, traditional representation. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10$ ,  $d_1 = 0.075$  cm,  $\epsilon_{r2} = 2$ ,  $d_2 = 0.15$  cm,  $\epsilon_{r3} = 1$ ,  $h = d_2/2$ ,  $z = 0.0$  cm,  $f = 1$  GHz.

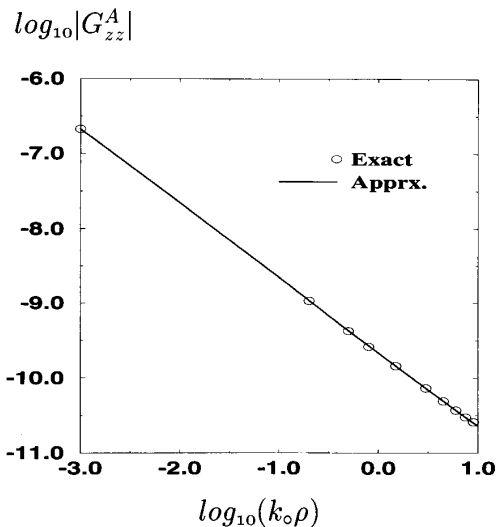


Fig. 8. Magnitude of the Green's function for the vector potential,  $G_{zz}^A$ , for a VED, traditional representation. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10$ ,  $d_1 = 0.075$  cm,  $\epsilon_{r2} = 2$ ,  $d_2 = 0.15$  cm,  $\epsilon_{r3} = 1$ ,  $h = d_2/2$ ,  $z = 0.0$  cm,  $f = 1$  GHz.

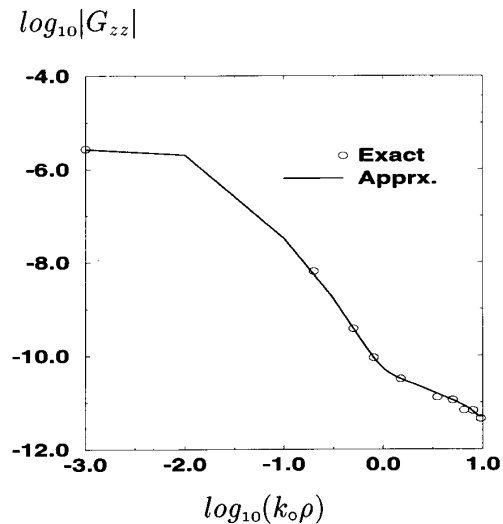


Fig. 10. Magnitude of the Green's function for the vector potential,  $G_{zz}^A$ , for a VED, alternative representation. Layer-0: PEC, Layer-3: half space,  $\epsilon_{r1} = 10$ ,  $d_1 = 0.075$  cm,  $\epsilon_{r2} = 2$ ,  $d_2 = 0.15$  cm,  $\epsilon_{r3} = 1$ ,  $h = d_2/2$ ,  $z = 0.0$  cm,  $f = 1$  GHz.

approximation of the spectral domain Green's functions by complex exponentials is performed by using the GPOF [17]. This technique is more robust and less noise sensitive [22] than the original and the least-square Prony's methods. The robustness of the technique comes from the fact that it utilizes the singular value decomposition technique as an intermediate step to extract the complex exponentials, through which the number of exponentials used in the approximation can be chosen as the number of the most significant singular values either automatically or interactively. Consequently, the number

of exponentials chosen for each Green's function in the examples given in this paper is different. For example, in the approximation of  $G_{xx}^A$  of Fig. 3, the number of exponentials, the number of samples, and the maximum sampled value  $T_o$  are chosen as 4 (including the direct term), 201, and 100, respectively, while the same parameters are chosen as 10 (including the direct term with no surface wave extraction), 401, and 100 for  $G_x^{qe}$  of Fig. 5. It should be noted that the numbers given above are strongly dependent on the parameters of the geometry and the source.

## IV. CONCLUSION

In this work, a complete set of closed-form, spatial domain Green's functions are provided in stratified media for general sources. The closed-form Green's functions are obtained using the GPOF method, which is more robust and less noise sensitive than the original Prony's and the least-square Prony's methods. In addition, the Green's functions are cast into a form to increase the numerical efficiency in the MoM applications. Numerical examples of the closed-form Green's functions are given for a multilayer medium. The approximate Green's functions are compared with the exact ones and very good agreement is observed.

## APPENDIX

## ALTERNATIVE FORM OF GREEN'S FUNCTION

Since the vector and scalar potentials are not uniquely defined in stratified media, different sets of Green's functions for the vector and scalar potentials are possible giving rise to many different MPIE formulations [18], [19]. Among these Green's functions, three useful choices referred as formulations A, B, and C, are given in [20]. In this paper, the formulation C is chosen as the alternative form of the Green's function and given here as

$$\begin{aligned} \tilde{G}_{A,F} = & (\hat{x}\hat{x} + \hat{y}\hat{y})G_{xx} + \hat{z}\hat{x}G_{zx} + \hat{z}\hat{y}G_{zy} \\ & + \hat{x}\hat{z}G_{xz} + \hat{y}\hat{z}G_{yz} + \hat{z}\hat{z}G_{zz} \end{aligned} \quad (30)$$

and  $G^{q_e,m}$  as the Green's function for the scalar potential for both horizontal and vertical dipoles, Fig. 1. Note that the difficulties encountered in the traditional formulation due to the difference between the scalar potentials of HED (HMD) and VED (VMD) are alleviated in this formulation. In the above form of the Green's function, the terms associated with the horizontal dipoles ( $G_{xx}^{A,F}$ ,  $G_{yy}^{A,F}$ ,  $G_{zx}^{A,F}$ ,  $G_{zy}^{A,F}$ ,  $G_x^{q_e,m}$ , and  $G_y^{q_e,m}$ ) remain the same as in the traditional form (2)–(7); two new entries,  $G_{xz}^{A,F}$  and  $G_{yz}^{A,F}$ , are introduced and  $G_{zz}^{A,F}$  is modified for the vertical dipoles. For an  $x$ -oriented HED and a  $z$ -oriented VED, the Green's function components used with the alternative form are adopted to the formulation given in Section II-A as

$$\begin{aligned} \tilde{G}_{zz}^A = & \frac{1}{jk_{z_i}} \left[ \frac{\mu_i}{2} e^{-jk_{z_i}|z|} + e^{-jk_{z_i}z} \left[ \frac{\mu_i}{2} A_v^e - \frac{k_{z_i}^2}{2\omega^2\epsilon_i} \right. \right. \\ & \cdot \left. \left. \left[ A_v^e - \frac{k_i^2 C_h^e - k_{z_i}^2 D_h^e}{k_\rho^2} \right] \right] + e^{jk_{z_i}z} \left[ \frac{\mu_i}{2} B_v^e + \frac{k_{z_i}^2}{2\omega^2\epsilon_i} \right. \right. \\ & \cdot \left. \left. \left[ -B_v^e + \frac{k_{z_i}^2 B_h^e + k_i^2 A_h^e}{k_\rho^2} \right] \right] \right] \end{aligned} \quad (31)$$

$$\begin{aligned} \tilde{G}_{xz}^A = & \frac{-j}{k_{z_i}} \frac{-k_x k_{z_i}}{2\omega^2\epsilon_i} \left[ e^{-jk_{z_i}z} \left[ A_v^e - \frac{k_i^2 C_h^e - k_{z_i}^2 D_h^e}{k_\rho^2} \right] \right. \\ & \left. + e^{jk_{z_i}z} \left[ -B_v^e + \frac{k_{z_i}^2 B_h^e + k_i^2 A_h^e}{k_\rho^2} \right] \right] \end{aligned} \quad (32)$$

where  $\tilde{G}_{zz}^A$ ,  $\tilde{G}_{xz}^A$  are the alternative Green's function components in the source layer and  $A_{h,v}^e$ ,  $B_{h,v}^e$ ,  $C_h^e$ ,  $D_h^e$  are given in the (12)–(17).

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## REFERENCES

- [1] N. K. Das and D. M. Pozar, "A spectral-domain Green's function for multilayer dielectric substrates with application to multilayer transmission lines," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-35, pp. 326–335, Mar. 1987.
- [2] F. Crog and D. M. Pozar, "Multifrequency operation of microstrip antennas using aperture coupled parallel resonators," *IEEE Trans. Antennas Propagat.*, vol. 40, pp. 1367–1374, Nov. 1992.
- [3] L. Barlatey, J. R. Mosig, and T. Sphicopoulos, "Analysis of stacked microstrip patches with mixed potential integral equation," *IEEE Trans. Antennas Propagat.*, vol. 18, pp. 608–615, May 1990.
- [4] A. N. Tulintseff, S. M. Ali, and J. A. Kong, "Input impedance of a probe fed stacked circular microstrip antenna," *IEEE Trans. Antennas Propagat.*, vol. 39, pp. 381–390, Mar. 1991.
- [5] K. R. Carver and J. W. Mink, "Microstrip antenna technology," *IEEE Trans. Antennas Propagat.*, vol. AP-29, pp. 2–24, Jan. 1981.
- [6] D. R. Jackson and N. G. Alexopoulos, "Gain enhancement methods for printed antennas," *IEEE Trans. Antennas Propagat.*, vol. AP-33, pp. 976–987, Sept. 1985.
- [7] C. I. G. Hsu, R. F. Harrington, K. A. Michalski, and D. Zheng, "Analysis of multiconductor transmission lines of arbitrary cross section in multilayered uniaxial media," *IEEE Trans. Microwave Theory Tech.*, vol. 41, pp. 70–78, Jan. 1993.
- [8] W. C. Chew, *Waves and Fields in Inhomogeneous Media*. New York: Van Nostrand Reinhold, 1990.
- [9] M. Gillick, I. D. Robertson, and J. S. Joshi, "Design analysis of novel coupling structures for multilayer MMIC's," *IEEE Trans. Microwave Theory Tech.*, vol. 41, pp. 346–349, Feb. 1993.
- [10] T. Tokumitsu, H. Hiraota, H. Makamoto, and T. Takenaka, "Multilayer MMIC using a  $3\ \mu\text{m} \times 3$ -layer dielectric film structure," in *IEEE MTT-S Int. Microwave Symp. Dig.*, 1990, pp. 831–834.
- [11] T. Hasegawa, S. Banba, H. Ogawa, and H. Nakamoto, "Characteristics of valley microstrip lines for use in multilayer MMIC's," *IEEE Microwave Guided Wave Lett.*, vol. 1, Oct. 1991.
- [12] M. I. Aksun and R. Mittra, "Choices of expansion and testing functions for the method of moments applied to a class of electromagnetic problems," *IEEE Trans. Microwave Theory Tech.*, vol. 41, pp. 503–509, Mar. 1993.
- [13] Y. L. Chow, J. J. Yang, and D. F. Fang, and G. E. Howard, "Closed form spatial Green's function for the thick substrate," *IEEE Trans. Microwave Theory Tech.*, vol. 39, pp. 588–592, Mar. 1991.
- [14] M. I. Aksun and R. Mittra, "Derivation of closed-form Green's functions for a general microstrip geometry," *IEEE Trans. Microwave Theory Tech.*, vol. 40, pp. 2055–2062, Nov. 1992.
- [15] ———, "Spurious radiation from microstrip interconnects," *IEEE Trans. Electromagn. Compat.*, vol. 35, pp. 148–158, May 1993.
- [16] S. L. Marple, *Digital Spectral Analysis with Applications*. Englewood Cliffs, NJ: Prentice Hall, 1987.
- [17] Y. Hua and T. K. Sarkar, "Generalized pencil-of-function method for extracting poles of an EM system from its transient response," *IEEE Trans. Antennas Propagat.*, vol. 37, pp. 229–234, Feb. 1989.
- [18] A. Erteza and B. K. Park, "Nonuniqueness of resolution of Hertz vector in presence of a boundary, and a horizontal dipole problem," *IEEE Trans. Antennas Propagat.*, vol. AP-17, pp. 376–378, May 1969.
- [19] K. A. Michalski, "On the scalar potential of a point charge associated with a time-harmonic dipole in a layered medium," *IEEE Trans. Antennas Propagat.*, vol. AP-35, pp. 1299–1301, Nov. 1987.
- [20] K. A. Michalski and D. Zheng, "Electromagnetic scattering and radiation by surfaces of arbitrary shape in layered media. Part I: Theory," *IEEE Trans. Antennas Propagat.*, vol. 38, pp. 335–344, Mar. 1990.
- [21] A. Sommerfeld, *Partial Differential Equations in Physics*. New York: Academic, 1949.
- [22] A. J. Mackay and A. McCowen, "An improved pencil-of-functions method and comparisons with traditional methods of pole extraction," *IEEE Trans. Antennas Propagat.*, vol. AP-35, pp. 435–441, Apr. 1987.



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